SIMULATION OF PHOTOVOLTAIC SYSTEM CONNECTED WITH FULL BRIDGE INVERTER USING MATLAB / SIMULINK

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Abstract

When sunlight shines on a PV cell, the absorbed light produces electricity. Though PV technologies use both direct and dispersed sunlight to create electricity, harnessing efficiency is 68% eventually against the claim of 85% by the various manufacturers worldwide. Power Electronics Interface are incorporated with Photovoltaic (PV) System to intensify the efficiency of the PV system and undoubtedly we have reached to the goalmouth. There are two stages where power electronics converter are used. First DC-DC converter stage in which lower level PV voltage is boosted-up at the required higher level; and second DC-AC inverter stage in which increased DC link voltage is efficiently converted into AC. Purpose of this paper is to elaborate the Full-Bridge inverter used in PV System and switching schemes adopted for the operation and to realize the best switching scheme. Simulation results are taken at various stages to visualize the effect of interface. For whole PV system simulation, PV module is connected to the converter system. The output of the Buck Boost converter is connected to the single-phase inverter and the inverter output is fed to the AC grid.

Key Words: H-Bridge Inverter, Switching Scheme, PV module, Converter, Simulink block-sets.

1. INTRODUCTION

For developing countries, providing energy to its stakeholders in an efficient and cost effective manner is a highly challenging task. In spite of significant harnessing of the fossil fuel reserves, the breach between supply and demand of energy is ever growing. One of the possible options to tie this breach is by making extensive use of solar power [4].

The need for a clean surroundings and the incessant increase in energy demand makes decentralized renewable energy production more and more substantial [4]. Photovoltaic generate electric power when illuminated by sunlight or artificial light. It directly convert the sun’s energy into electricity which can be easily transported and converted to other forms for the benefit of society [16]. Though PV technologies use both direct and dispersed sunlight to create electricity, harnessing efficiency is 68% eventually against the claim of 85% by the various manufacturers worldwide. Power Electronics Interface are incorporated with PV System to intensify the efficiency of the PV system and undoubtedly we have reached to the goalmouth.

Power Electronics is the field of engineering which deals with the use of electronics for the conversion, control and conditioning of bulk electrical power. It also plays an important role in the solar system [11], [12], [13], [15]. There are two stage where power electronics converter are used. First DC-DC converter stage in which lower level PV voltage is stepped-up at the required higher level [24]; and second DC-AC inverter stage in which boosted DC link voltage is converted into AC [17], [18]. If Maximum Power Point Tracking (MPPT) is accountable for optimizing the efficiency of the photovoltaic system, power electronics interface is the solver. The power losses incurred at the converter stage is reimbursed at inverter stage. This proposed PV module gives the maximum power, voltage and current independent of the load.

2. DC-AC INVERTER STAGE

The power produced by a PV module is in the form of direct current. Conversion of direct current to alternating current required by many common appliances and for grid-connection is realized with an inverter. Inverter is basically an interface between photovoltaic cell and AC grids. There are several inverter topologies but output current distortion and efficiency are the two key parameters for the selection of inverters. Power inverters produce one of three different types of wave output [1] [2] [3] [4].

- Square Wave
- Modified Sine Wave
- Pure Sine Wave

Based on their operation the inverters can be broadly classified into

- Voltage source inverter (VSI)
- Current source inverter (CSI)

The output voltage waveform of Voltage Source Inverter are independently controlled and mostly remain unaffected by the load. Due to this phenomenal behavior, the VSI have many industrial applications such as adjustable speed drives and also in Power system for FACTS (Flexible AC
Transmission System) [21]. The voltage source inverter possesses a capacitor in parallel with the DC input [2], [5].

The output current waveform of Current Source Inverter are independently controlled and mostly remain unaffected by the load. Such Inverters are broadly used in medium voltage industrial applications where high quality waveform is mandatory. The current source inverters possess an inductor in series with the input. The converter topology are shown in Fig 1(a) and (b). Full - Bridge inverter are widely used in Photovoltaic system [26]. Therefore here we will discuss the full bridge voltage source inverter and its switching schemes.

2.1 Full Bridge Inverter

Fig 2 the Full bridge inverter consists of two parallel strings contributing two switching power devices in series with anti-parallelled diodes. Variation of duty cycle of the PWM signal offers a voltages across the load in a specific pattern which appear as the load as AC signal. A pure sin wave is attained after passing the signal through a low pass filter [25]. The pattern at which the duty cycle of a PWM signal varies can be realized using simple analogue components or a digital microcontroller [20]. Either of the two basic topologies generate sinusoidal PWM that controls the output of the inverter [1]. The full bridge converter can be used to generate two different PWM pulse trains depending on the switching scheme implemented. The two schemes are called bipolar and unipolar switching [6], [7], [13], [15], [22].

2.1.1 Bipolar Switching Scheme

A full bridge converter which uses a bipolar switching scheme is called a two level converter.

Table 1: switching states for bipolar switching scheme for a full bridge converter

<table>
<thead>
<tr>
<th>Switching State</th>
<th>Switches On</th>
<th>Switches Off</th>
<th>Pulsed Output Voltage (V_j)</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>S_1, S_4</td>
<td>S_2, S_3</td>
<td>+V_{dc}</td>
</tr>
<tr>
<td>2</td>
<td>S_2, S_3</td>
<td>S_1, S_4</td>
<td>-V_{dc}</td>
</tr>
</tbody>
</table>

With a bipolar switching scheme, the full bridge converter only has two switching states as the junction voltage (V_j) switches from +V_{dc} to -V_{dc} using PWM. When used as a two level converter the full bridge converter uses the switching scheme outlined in Table I the resulting pulsed output is displayed in Fig 3. Figure 4 displays the carrying current paths at each switching state [6].
2.1.2 Unipolar Switching Scheme

A full bridge converter which uses unipolar switching is called a three level converter. A unipolar switching scheme is where the output of the converter \( V_d \) switches between \( +V_{dc} \) and zero during the positive half wave and between \( -V_{dc} \) and zero during the negative half wave.

The three unipolar switching schemes implemented by full bridge topologies that schemes and associated switching states and orders are presented in Table 2, 3 & 4.

**Table -2: Switching States for Standard Unipolar Switching Scheme for a Full Bridge Converter**

<table>
<thead>
<tr>
<th>Switching state</th>
<th>Switch on</th>
<th>Switch off</th>
<th>Junction voltage</th>
<th>Half wave</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>S1, S4</td>
<td>S2, S3</td>
<td>(+V_{dc})</td>
<td>Positive</td>
</tr>
<tr>
<td>2</td>
<td>S1, S3</td>
<td>S2, S4</td>
<td>0</td>
<td>Positive</td>
</tr>
<tr>
<td>3</td>
<td>S1, S4</td>
<td>S2, S3</td>
<td>(-V_{dc})</td>
<td>Positive</td>
</tr>
<tr>
<td>4</td>
<td>S2, S4</td>
<td>S1, S3</td>
<td>0</td>
<td>Positive</td>
</tr>
<tr>
<td>5</td>
<td>S2, S3</td>
<td>S1, S4</td>
<td>(+V_{dc})</td>
<td>Negative</td>
</tr>
<tr>
<td>6</td>
<td>S1, S3</td>
<td>S2, S4</td>
<td>0</td>
<td>Negative</td>
</tr>
<tr>
<td>7</td>
<td>S2, S3</td>
<td>S1, S4</td>
<td>(-V_{dc})</td>
<td>Negative</td>
</tr>
<tr>
<td>8</td>
<td>S2, S4</td>
<td>S1, S3</td>
<td>0</td>
<td>Negative</td>
</tr>
</tbody>
</table>

**Table - 3: Switching States for One Phase Chopping Unipolar Switching Scheme (type –a) for a full bridge converter**

<table>
<thead>
<tr>
<th>Switching state</th>
<th>Switch on</th>
<th>Switch off</th>
<th>Junction voltage</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>S1, S4</td>
<td>S2, S3</td>
<td>(+V_{dc})</td>
</tr>
<tr>
<td>2</td>
<td>S2, S3</td>
<td>S1, S4</td>
<td>0</td>
</tr>
<tr>
<td>3</td>
<td>S2, S4</td>
<td>S1, S3</td>
<td>(-V_{dc})</td>
</tr>
<tr>
<td>4</td>
<td>S2, S4</td>
<td>S1, S3</td>
<td>0</td>
</tr>
</tbody>
</table>

Unlike bipolar switching, unipolar switching requires at least three different switching states as the junction voltage \( V_d \) can be either \(+V_{dc}, -V_{dc}\) or zero although most inverters implement four switching states by having a different switching combination to create the zero junction voltage for each half wave. This is to evenly distribute the use of switches making heating symmetrical and thereby reducing losses [6], [19].

A number of different unipolar switching schemes exist for full bridge topologies with associated advantages and disadvantages regarding PV array voltage, switching losses and complexity of control signal generation [23].

The main difference between the three presented unipolar switching schemes is the implementation of different freewheeling states during each half wave. While both one phase chopping methods operate similarly with the only difference being that type B implements a different switching state for the freewheeling state of each half wave, the standard method is subtly different because both freewheeling states occur in the same half wave [6].

The resulting pulsed output of the one phase chopping scheme (type B) is displayed in Fig 5. The associated current paths are displayed in Fig 6 [6].

One of the main advantages of executing a unipolar switching scheme compared to a bipolar scheme is that the switching losses are significantly reduced due to the allied voltage drop of switching from one state to another. One drawback of executing a unipolar switching scheme is that, there are higher associated harmonic content in the output current round the zero crossing (particularly at lower power levels).
The Simulink block diagram of the current control loop is shown in Fig 8. First of all an external voltage signal corresponding to the grid voltage is fed into a discrete single-phase phase-locked-loop (PLL). The gain at the input of PLL is recycled to normalize the actual voltage signal. The output of the PLL block produces a reference sine wave. This Sine wave together with the gain block is used to create the reference current.

Fig – 8: Simulink model of the grid current control scheme

Table – 5: Simulation Parameter for Converter

<table>
<thead>
<tr>
<th>Description</th>
<th>Rating</th>
</tr>
</thead>
<tbody>
<tr>
<td>Inductor</td>
<td>85µH</td>
</tr>
<tr>
<td>Capacitor</td>
<td>300µF</td>
</tr>
<tr>
<td>Resistive Load</td>
<td>100Ω</td>
</tr>
<tr>
<td>Switching Frequency</td>
<td>10KHz</td>
</tr>
</tbody>
</table>

Saturation block is used to ensure the controlled signal do not saturate. This saturated signal is given to the pulse generator which provides the gate pulses for the Inverter.
3.1 Simulation Results

Fig 9 shows the output current from the photovoltaic module. From this result we can see that photovoltaic current is gradually decreases from short circuit current 5.2A to the 5.17A and after this it remain constant.

Fig 10 shows the output voltage from the photovoltaic module. From this result we can see that photovoltaic voltage increases from zero to 32.8 volt. After this it remains constant.

Fig 11 shows the output power from the photovoltaic module. Maximum power output from the photovoltaic module is 169.5W which is almost equal to $P_{mpp}$ given into the data sheet of the photovoltaic module.

Fig 12 shows the buck boost converter output voltage. Input voltage from the photovoltaic module 32.8 is boosted up to 200 volt.

Fig 13 shows the output power from the photovoltaic module. Maximum power output from the photovoltaic module is 169.5W which is almost equal to $P_{mpp}$ given into the data sheet of the photovoltaic module.

Fig 14 shows the voltage across Buck boost converter switch.

Fig 15 shows the buck boost converter inductor current which is discontinuous.

Buck boost converter output voltage is converted into AC by using Full bridge inverter. Fig 16 is the output voltage of inverter without filter.
Fig 17 and Fig 18 shows the filtered output current and voltage of the inverter. The PV module gives the maximum power, voltage and current independent of the load. 32.8 volt photovoltaic voltage is boosted up to 200 volts without any loss of power.

Fig 19 shows the FFT analysis of the output current of inverter for one cycle. Here THD is equal to 9.02%.

Fig 20 shows the FFT analysis of the output current of inverter for three cycles. Here THD is equal to 7.18 %.

4. CONCLUSIONS

- Photovoltaic current is gradually decreases from short circuit current 5.2A to the 5.17A and after this it remain constant [Fig. 9].
- Photovoltaic voltage increases from zero to 32.8 volt & after this value it remain constant [Fig. 10].
- Maximum power output from the photovoltaic module is 169.5W which is almost equal to $P_{app}$ given into the data-sheet [Fig. 11].
- Buck-boost converter output voltage is boosted up from 32.8 volt to 200 volt [Fig.12].
- Switching losses are significantly reduced due to the allied voltage drop of switching from one state to another in unipolar switching scheme.
- The PV module gives the maximum power, voltage and current independent of the load.
- The power loses incurred at the converter stage is reimbursed at inverter stage.
- Thus the overall efficiency and reliability of the photovoltaic system is improved.

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BIOGRAPHIES

Prof Anwarul M. Haque was born in India, Bihar, in 1967. He received his B.E (Power Electronics) in 1989 and M. Tech (Electrical – Power System) in 2012 respectively from Saurashtra University and Manav Bharti University. Currently he is working with VGE, Ahmedabad, Gujarat and pursuing his PhD (Electrical Engineering) from Jodhpur National University, Jodhpur, Rajasthan since July 2012.

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Prof. Nagal is a Member of INSTITUTE OF ENGINEERS (INDIA) and INTERNATIONAL ASSOCIATION OF ENGINEERS (HONGKONG).

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